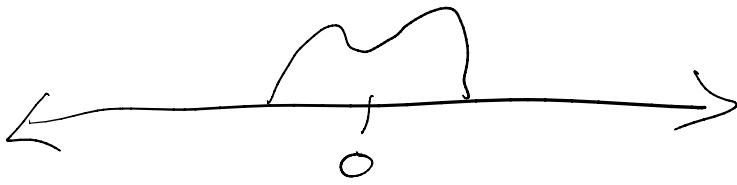


$m(t)$

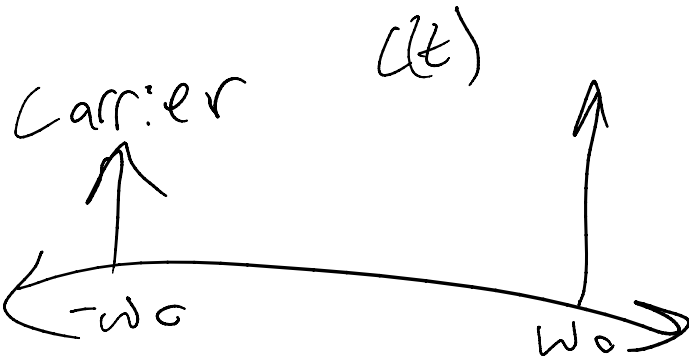
in Matlab



$m[n]$  sampled at  $f_s^{(a)}$

$\sim 8 \text{ kHz}$

$\sim 40 \text{ kHz}$



$f_c > 100 \text{ kHz}$

$c[k]$  sampled at  $f_s^{(1)}$

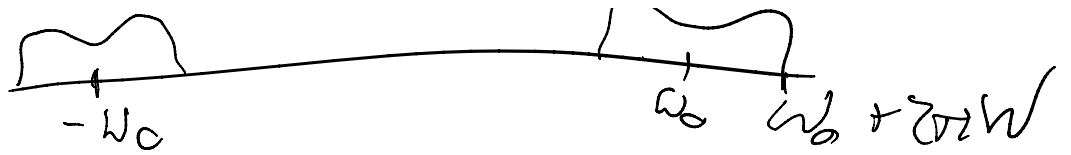
$f_s^{(1)}$  to represent  $c(t)$  must be  $\geq 2f_0$  ← Nyquist sampling theorem

so to take  $m(t), c(t) \rightarrow$  Matlab

$m[k], c[k]$  at  $f_s^{(1)}$

and





So really  $f_s^{(1)} > 2f_c + 2w$   
 $\sim 2f_c + 40 \text{ kHz}$



mag plot - semilog  
 phase - unwrap

# Prob Review

Tuesday, September 15, 2020 6:08 PM

Sample space  $\Omega$  - set of outcomes of a random exp.

Prob. fnch is  $P: \mathcal{F}(\Omega) \rightarrow \mathbb{R}$  satisfying

1)  $0 \leq P(E) \leq 1 \quad \forall E \subset \Omega$

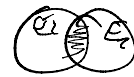
2)  $P(\Omega) = 1$

3) if  $E_1, \dots, E_k$  are mutually exclusive  
( $E_i \cap E_j = \emptyset \quad \forall i, j$ ) then  $P(\bigcup_i E_i) = \sum_i P(E_i)$

if a sample space is countable  $\rightarrow$  discrete  
uncountable  $\rightarrow$  cts

$$P(E^c) = 1 - P(E)$$

$$P(\emptyset) = 0$$

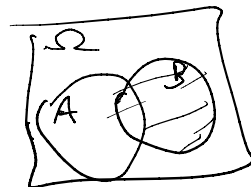


$$P(E_1 \cup E_2) = P(E_1) + P(E_2) - P(E_1 \cap E_2)$$

if  $E_1 \subset E_2$ ,  $P(E_1) \leq P(E_2)$

Def  $A, B$  independent iff  $P(A \cap B) = P(A)P(B)$

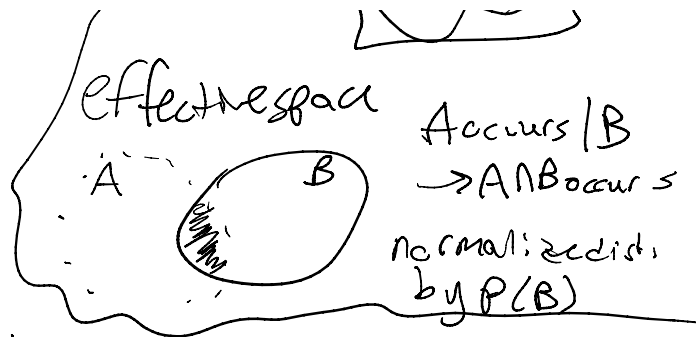
Def  $P(A|B) = \frac{P(A \cap B)}{P(B)}$



ref. we read

and

$P(X|B)$   
satisfies the  
def of a prob.



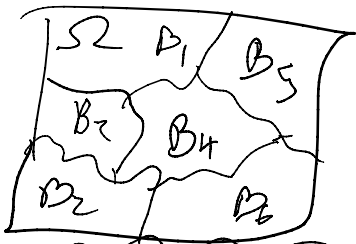
fn on invar: ablx w/ sample space  $\Omega$  if  $P(B) \neq 0$

If events are indep,  $P(B|A) = \frac{P(A \cap B)}{P(A)} = \frac{P(A)P(B)}{P(A)} = P(B)$   
 $P(A|B) = P(A)$

Def. A partition of the sample space  $\Omega$  is  
 a set of events  $B_1, \dots, B_n$  satisfying

1)  $\bigcup_{i=1}^n B_i = \Omega$  (exhaustive)

2)  $B_i \cap B_j = \emptyset \quad \forall i \neq j$  (mutual exclusivity)



Total Probability Theorem: If  $\{B_i\}_{i=1}^n$  is a partition  
 of  $\Omega$ , then for any  $A \subset \Omega$ , I can write

$$P(A) = \sum_{i=1}^n P(B_i) P(A|B_i)$$

Proof.  $P(B_i)P(A|B_i) = P(A \cap B_i) \quad \forall i$

$$\text{so } \sum_i P(B_i)P(A|B_i) = \sum_i P(A \cap B_i)$$

if  $B_i$ 's are m.e., then  $A \cap B_i$ 's are m.p.

so by Prob. def (3),

$$= P\left(\bigcup_i (A \cap B_i)\right)$$

$$\text{but } \bigcup_i B_i = \Omega, \text{ so } = P(A) //$$

Bayes' Theorem  $P(B|A) = \frac{P(A|B)P(B)}{P(A)}$

Proof  $P(B|A) = \frac{P(B \cap A)}{P(A)}$ ,  $P(B \cap A) = P(A|B)P(B)$

$$P(B|A) = \frac{P(A|B)P(B)}{P(A)} //$$

an alternate (equivalent) statement:

if  $\{B_i\}_{i=1}^n$  is a partition

$$P(B_k|A) = \frac{P(A|B_k)P(B_k)}{\sum_i P(A|B_i)P(B_i)}$$

$P(B_i)$  "priors"

$P(B_i)$  "priors"

$P(B_k|A)$  "posterior"

Ex. Covid19 antibody test has false neg. rate of 4%, false pos. rate of 10%. You take the test and you test positive.

What is the prob. that you have the antibodies given 20% of essential workers like you have them?

$A$  = having antibodies,  $B$  = Testing positive

$$P(A|B) = \frac{P(B|A)P(A)}{P(B|A)P(A) + P(B|A^c)P(A^c)} \quad (\text{partition: } \{A, A^c\})$$

$$= \frac{(1 - P(B^c|A)) (.2)}{(1 - P(B^c|A)) (.2) + (.1) (.8)}, \quad P(B^c|A) = .04$$

$$= \frac{(.96)(.2)}{(.96)(.2) + .08} \approx 71\%$$

A	A <sup>c</sup>
T <sup>+</sup>	
T <sup>-</sup>	

Random Variables are functions from sample space  $\Omega$

$$X: \Omega \rightarrow \mathbb{C} \text{ (or } \mathbb{R})$$

well,  $\Omega$  may not be numerical

well,  $\Omega$  may not be numerical

$$\Omega = \{\text{Heads, Tails}\}$$

$$X(H) = 0 \quad Y(H) = -1$$

$$X(T) = 1 \quad Y(T) = 1$$

Same exp., 2 vars, 2 different means  
Use r.v.s as a means of mapping  $\Omega$  to a space where  
we can "do math" and define things like distributions  
or Expectations

we call  $\text{range}(X)$  "the space" of  $X$ ,  $S$ .

if  $S$  countable, then  $X$  is discrete

uncountable, then  $X$  is Cts

---

Def. P.M.F. for a.r.v.  $X$  with space  $S$  as

$$f_X : S \rightarrow [0, 1]$$

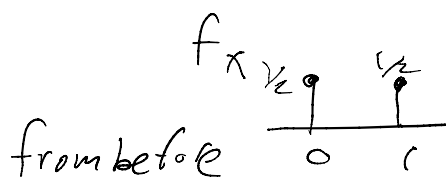
$$f_X(x_i) = P(X = x_i)$$

$$= P(X^{-1}(x_i))$$

↑ subset of  $\Omega$

$$\{\omega \mid X(\omega) = x_i\} \subset \Omega$$

$$\sum f_X(x_i) = 1, \text{ and def. } f_X(x) = 0 \text{ if } x \notin S$$



$$\sum_{x_i \in S} f_X(x_i) = 1, \text{ and def. } f_X(x) = 0 \text{ if } x \notin S$$

cts case: P.D.F. of cts r.v.  $X$  space  $S$

$$f: S \rightarrow \mathbb{R}^+, \quad \int_I f_X(y) dy = P(X \in I) \\ := P(X^{-1}(I))$$

$$\int_{\mathbb{R}} f_X(y) dy = 1$$

CDF:  $F_X(y) = \int_{-\infty}^y f_X(z) dz = P(X \leq y)$

(can be def. for discrete r.v.s)

a)  $0 \leq F_X(x) \leq 1$ , b)  $F_X$  is non decreasing

c)  $\lim_{x \rightarrow -\infty} F_X(x) = 0$ ,  $\lim_{x \rightarrow \infty} F_X(x) = 1$

d)  $P(X \in (a, b)) = F_X(b) - F_X(a)$

e)  $f_X(x) = \left. \frac{d}{dy} F_X(y) \right|_{y=x}$



$$f'_x(x) = \left. \frac{d}{dy} f_x(y) \right|_{y=x}$$

Some important r.v.s

1. Bernoulli experiment is a "success" with prob.  $p$   
"failure" with prob.  $1-p$

good model for some binary channels, bitwise

2. Binomial I perform  $n$  Bernoulli trials (indep), counts

# successes  $P(X=k) = \binom{n}{k} p^k (1-p)^{n-k}$

"how many bit errors in an  $n$ -bit communication"

3. Uniform  $X = U(a, b)$

$$f_x(x) = \begin{cases} \frac{1}{b-a}, & a < x < b \\ 0, & \text{else} \end{cases}$$



4. Normal (Gaussian)  $N(\mu, \sigma^2)$

$$f_x(x) = \frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{(x-\mu)^2}{2\sigma^2}}$$

Common model for thermal noise  
(most types of noise)

Q function:  $Q(x) = \int_x^{\infty} f_x(x) dx$  where

Q function:  $Q(x) = \int_x f_X(x) dx$  where  $X = N(0, 1)$

in Matlab, its inverse is too,  $Q(0) = 1/2$

if  $X = N(0, 1)$ , then

not squared!

$$N(\mu, \sigma^2) = \sigma X + \mu$$

generated by random multiply by

Expectation

disc:  $E[X] = \sum_i x_i f_X(x_i)$

cts:  $E[X] = \int_{-\infty}^{\infty} x f_X(x) dx$

$$E[X] = \mu_X, \quad E[(X - \mu)^2] = \sigma_X^2 \leftarrow \text{variance}$$

$\uparrow$   
mean

$\sigma_X$  "std. deviation"

Expectation is linear!

$$E[aX + b] = aE[X] + b$$

if  $Y = g(X)$ , then  $E[g(X)] = \int g(x) f_X(x)$

if  $Y = g(X)$ , then  $E[g(X)] = \sum g(x) f_X(x)$   
or  $\int g(x) f_X(x) dx$

$N(\mu, \sigma^2)$  has mean  $\mu$ , var.  $\sigma^2$

---

Multiple R.V.

joint pmf.  $f_{X,Y}(x,y) = P(X=x \text{ and } Y=y)$

marginal pmf  $f_X(x) = \sum_{y \in S_Y} f_{X,Y}(x,y)$

$$f_X(x) = \int_{\mathbb{R}} f_{X,Y}(x,y) dy$$

joint CDF  $F_{X,Y}(x,y) = \int_{-\infty}^y \int_{-\infty}^x f_{X,Y}(u,v) du dv = P(X \leq x \text{ and } Y \leq y)$

---

two random variables are indep.

$$f_{X,Y}(x,y) = f_X(x) f_Y(y)$$

---

Conditional dist

$$f_{X|Y}(x|y) = \begin{cases} \frac{f_{X,Y}(x,y)}{f_Y(y)}, & f_Y(y) \neq 0 \\ 0, & \text{else} \end{cases}$$

$$x/y = \begin{cases} 1 & \text{if } x=y \\ 0 & \text{else} \end{cases}$$

if  $X, Y$  indep. then  $f_{X,Y}(x/y) = f_X(x)$

Correlation b/w  $X$  and  $Y$ :  $\rho_{X,Y} = E[XY]$

Covariance:  $\sigma_{X,Y} = E[(X-\mu_X)(Y-\mu_Y)]$   
 $= E[XY] - E[X]E[Y]$

Correlation Coefficient:  $\rho_{X,Y} = \frac{\sigma_{X,Y}}{\sigma_X \sigma_Y}$

We say  $X, Y$  are orthogonal if  $E[XY] = 0$   
 that is - if  $\sigma_{X,Y} = 0$

We say  $X, Y$  are uncorrelated if  $\sigma_{X,Y} = 0 \Rightarrow \rho_{X,Y} = 0$

if  $X, Y$  indep:  $\sigma_{X,Y} = E[XY] - E[X]E[Y]$   
 $= E[X]E[Y] - E[X]E[Y] = 0$

indep  $\Rightarrow$  uncorrelated

converse is false

Central limit Theorem

If  $X_1, \dots, X_n$  are indep., identically distributed  
 $\overset{n}{\uparrow}$  . . .

$$\frac{1}{n} \sum_{i=1}^n X_i \longrightarrow N(\mu, \sigma^2/n) \text{ as } n \uparrow$$

where  $\mu, \sigma^2$  are mean, var of  $X_i$

# Random Process

Tuesday, September 15, 2020 7:37 PM

A random process is a generalization of the r.v.

It is a set of possible functions of time rather than values

Ex.

Suppose  $X(t) = A \cos(\omega_0 t + \theta)$ ,  $\theta = U(0, 2\pi)$

↑  
random process

at any point  $t_0$  in time,  $X(t_0)$  is a random variable

$$A \cos(\omega_0 t_0 + \theta)$$

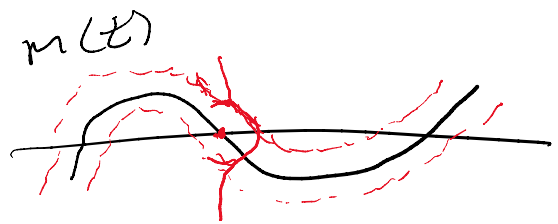
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A random process is a function of a real variable  $t$  (time) which is a random variable at each value of  $t$ .

Ex.

$$X(t) = m(t) + N(0, \sigma^2)$$

$$= N(m(t), \sigma^2)$$



---

Discrete time random process = at each instance in discrete time is a r.v. =  $\{x_i, i = 0, \infty\}$  where each  $x_i$  is an r.v.

is a r.v.  $\equiv \sum_{n=0}^{\infty} X_n$  where each  $X_n$  is an r.v.

---

# Expectation

Tuesday, September 15, 2020 7:49 PM

say  $X(t)$  is a r. p.

So at time  $t_0$ ,  $E[X(t_0)]$  is well-defined if  $X(t)$  has a r.v.

just a random variable

So def.  $m_x(t) = E[X(t)]$

or  $m_x(t_0) = E[X(t_0)] \quad \forall t_0 \in \mathbb{R}$

def.  $\sigma_x^2(t) = E[(X(t) - m_x(t))^2]$

Ex.  $X(t) = A \cos(\omega t + \Theta)$ ,  $\Theta = U(0, 2\pi)$ ,

$$m_x(t) = \int_0^{2\pi} A \cos(\omega t + \theta) \left(\frac{1}{2\pi}\right) d\theta = 0 \quad \text{const!}$$



Def The auto correlation function of r.p.  $X(t)$  to be

$$R_x(t_1, t_2) = E[X(t_1)X(t_2)]$$

$$= \iint x_1 x_2 f_{X(t_1), X(t_2)}(x_1, x_2) dx_1 dx_2$$

just  $\int_{X(t_1)X(t_2)}$

Ex.  $X(t) = A \cos(\omega t + \Theta)$   $\Theta = U(0, 2\pi)$

$$R_x(t_1, t_2) = E[(A \cos(\omega t_1 + \Theta)) (A \cos(\omega t_2 + \Theta))] \\ = \frac{A^2}{2} E[\cos(\omega(t_1 - t_2)) + \cos(\omega t_1 + \omega t_2 + 2\Theta)]$$

$$= \frac{A^2}{2} \cos(\omega(t_1 - t_2))$$

depends only on  $t_1 - t_2$  } The time difference

Def A process  $X(t)$  is Wide Sense Stationary (WSS) if

1)  $\mu_X(t)$  is indep. of  $t$  and

2)  $R_X(t_1, t_2) = R_X(\tau)$  is a function of only

$$t_1, t_2 = \tau$$

---

So  $A \cos(\omega t + \phi)$  was WSS!

because for WSS,  $R_X(t_1, t_2)$  depends only on delay, we have

$$\left. \begin{aligned} R_X(t_1, t_2) &= R_X(t_2, t_1) \\ R_X(\tau) &= R_X(-\tau) \end{aligned} \right\} \begin{array}{l} \text{for WSS} \\ R_X \text{ is even in } \tau \end{array}$$

Def A process is cyclostationary if  $\mu_X$  and  $R_X(t_1, t_2) = R_X(t + \tau, t)$  are periodic with same period  $T_0$

$$\left. \begin{aligned} \text{i.e. } \mu_X(t + T_0) &= \mu_X(t) \\ R_X(t + \tau + T_0, t + T_0) &= R_X(t + \tau, t) \end{aligned} \right\}$$

none writes like this  $\rightarrow R_X(t_1 + T_0, t_2 + T_0) = R_X(t_1, t_2)$

Ex.  $X(t) = A \cos \omega t + AN(t)$  where  $N(t)$  is an indep. sample of  $N(0, 1)$  at each  $t$

$$\mu_X(t) = E[A \cos \omega t + AN(t)]$$

$$= A \cos \omega t, \text{ periodic w/ period } 2\pi/\omega$$

$$R_X(t+\tau, t) = E[(A \cos(\omega t + \omega\tau) + AN(t+\tau))(A \cos(\omega t) + AN(t))]$$

$$= A^2 E[\cos(\omega(t+\tau)) \cos \omega t + N(t) \cos(\omega(t+\tau)) + N(t+\tau) \cos \omega t + N(t)N(t+\tau)]$$

$$= A^2 (\cos(\omega(t+\tau)) \cos \omega t + 0 + 0 + E[N(t)] E[N(t+\tau)])$$

$$= A^2 \cos \omega t \cos(\omega(t+\tau)) \text{ periodic w/ period } 2\pi/\omega$$

X cyclostationary

For  $X$  cyclostationary, we define

$$\overline{R_X}(\tau) = \frac{1}{T_c} \int_0^{T_c} R_X(t+\tau, t) dt$$

// average autocorrelation function of only the delay

---

Ex.  $A(\cos \omega t + N(t))$

$$R_X(t+\tau, t) = A^2 \cos \omega t \cos(\omega(t+\tau)), \quad T = \frac{2\pi}{\omega}$$

$$\begin{aligned} \overline{R_X}(\tau) &= \frac{A^2}{T} \int_0^T \cos \omega t \cos(\omega(t+\tau)) dt \\ &= \frac{A^2}{2T} \int_0^T \cos(\omega \tau) + \cancel{\cos(2\omega t + \omega \tau)} dt \\ &= \frac{A^2}{2} \cos \omega \tau \end{aligned}$$

Def  $X(t), Y(t)$  indep. if  $\forall m, n \in \mathbb{N}$

$\forall t_1, \dots, t_m, \tau_1, \dots, \tau_n \in \mathbb{R}$  we have

$(X(t_1), \dots, X(t_m))$  and  $(Y(\tau_1), \dots, Y(\tau_n))$  are indep

---

$X(t), Y(t)$  indep  $\Rightarrow X(t_1), Y(t_2)$  are indep as r.v.s

Def Cross correlation b/w  $X(t), Y(t)$  by

$$R_{xy}(t_1, t_2) = E[X(t_1)Y(t_2)]$$

indep  $\Rightarrow$  uncorrelated

Def  $X(t), Y(t)$  are jointly WSS if  $X, Y$

are each WSS and

$R_{xy}(t_1, t_2)$  dep. only on  $t_1 - t_2 = \tau$

So we have  $R_{xy}(\tau)$

Let's get back to signals/systems

LTI system w/ impulse resp.  $h$

input r.f.  $X(t)$

$$X(t) \rightarrow [h] \rightarrow Y(t) = X(t) * h$$

Claim If  $X$  is WSS, then  $Y$  and  $X$  are jointly WSS

and 
$$\mu_Y = \mu_X \int_{-\infty}^{\infty} h(t) dt$$

$$R_{XY}(\tau) = R_X(\tau) * h(-\tau)$$

$$R_Y(\tau) = R_X(\tau) * h(\tau) * h(-\tau)$$

Proof (see text or ... later)



Ex.  $h(t) = \delta(t - t_0)$  delay

$x \rightarrow [h] \rightarrow y$

$$y(t) = x(t - t_0)$$

now - input  $x(t)$  WSS

$$\mu_y = \mu_x \int_{-\infty}^{\infty} \delta(t - t_0) dt = \mu_x$$

$$R_{xy} = R_x(\tau) * \delta(-\tau - t_0) = R_x(\tau + t_0)$$

$$R_y = R_x(\tau) * \delta(-\tau - t_0) * \delta(\tau - t_0)$$

$$= R_x(\tau + t_0) * \delta(\tau - t_0)$$

$$= R_x(\tau + t_0 - t_0) = R_x(\tau)$$

Ex.  $h(z) = \frac{1}{\pi z}$  ← H.T.

$$M_Y = M_X \int_{-\infty}^{\infty} \frac{1}{\pi z} dz = 0 \quad \text{(odd function)}$$

$$R_{XY}(\tau) = R_X(\tau) * \frac{1}{-\pi z} = -\hat{R}_X(\tau)$$

$$R_Y(\tau) = R_X(\tau) * \frac{1}{\pi z} * \frac{1}{\pi z} = -\hat{\hat{R}}_X(\tau) \\ = R_X(\tau)$$

## Freq. Domain?

Define an analogue to F.T.

Power Spectral Density gives the power of a random process at diff. frequencies

Def.  $S_X(\omega) = E[|X(\omega)|^2]$

THM (Wiener - Khinchin)

For a WSS process, the P.S.D. is given by

$$S_X(\omega) = \mathcal{F}\{R_X(\tau)\}(\omega)$$

For a cyclostationary process,

$$S_X(\omega) = \mathcal{F}\{\bar{R}_X(\tau)\}(\omega)$$

This is the def. for random signals

The power content of a process is def. by

$$P_x = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_x(\omega) d\omega$$

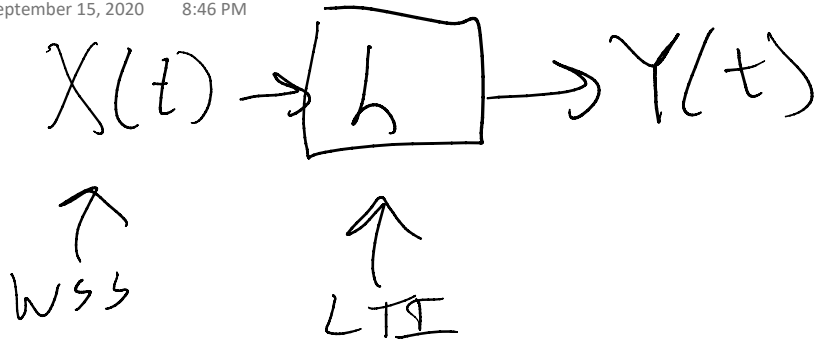
if a process is WSS, then

$$R_x = \mathcal{F}^{-1} \{ S_x(\omega) \}$$

$$\text{so } R_x(\tau) = \int S_x(\omega) e^{j\omega\tau} d\omega$$

$$\text{so } R_x(0) = \int S_x(\omega) d\omega = P_x$$

$$\boxed{\text{for WSS, } P_x = R_x(0)}$$



$$\mu_Y = \mu_X \int h$$

$$R_Y = R_X * h(-\tau) * h(\tau)$$

↓ FT both sides

$$S_Y(\omega) = |H(\omega)|^2 S_X(\omega) \leftarrow \text{use this a lot}$$

$$\mathcal{F}\{R_X\} = \mathcal{F}\{S_{XX}\}$$

$$\mu_Y = \mu_X H(\omega)$$